

## The Uncertainty Inequality in Signal Processing

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Communication engineers interpret the Fourier transform as decomposing an information-carrying signal  $f(x)$ , where  $x$  represents time, into a superposition of pure sinusoidal “tones” having frequencies represented by a real variable. In fact, engineers usually think about the resulting “frequency domain” representation as much as or more than the “time domain” representation (that is, the signal itself)! A fundamental fact of signal processing is that the narrower a signal is in the time domain, the broader it is in the frequency domain. Conversely, the narrower a signal is in the frequency domain, the broader it is in the time domain. This effect is important because in practice, a signal must be sent in a limited interval of time and using a limited interval, or “band,” of frequencies. In this project, we describe and investigate this trade-off between duration and bandwidth both qualitatively and quantitatively. The results of our investigation will support a commonly quoted rule of thumb: The number of different signals that can be sent in a certain duration of time using a certain band of frequencies is proportional to the product of the time duration and the width of the frequency band.

### Related Problems

We will use the complex form of the Fourier transform and inverse Fourier transform given in (5) and (6) of Section 15.4. We will use the notation  $\hat{f}(\alpha)$  to denote the Fourier transform of a function  $f(x)$  in a compact way that makes its dependence on  $f$  explicit—that is,  $\hat{f}(\alpha) = \mathcal{F}\{f(x)\}$ . We take  $f$  to be a real-valued function, and we warm up by noting two simple properties that  $\hat{f}$  enjoys.

1. Show that if  $\alpha > 0$ , then  $\hat{f}(-\alpha) = \overline{\hat{f}(\alpha)}$ . So for any  $\alpha$ ,  $|\hat{f}(-\alpha)| = |\hat{f}(\alpha)|$ . (Here the notations  $\bar{z}$  and  $|z|$  represent the conjugate and the modulus of a complex number  $z$ , respectively.)
2. If  $k$  is a real number, let  $f_k(x) = f(x - k)$ . Show that

$$\hat{f}_k(\alpha) = e^{iak} \hat{f}(\alpha)$$

So shifting a signal in time does not affect the values of  $|\hat{f}(\alpha)|$  in the frequency domain.

Keeping these facts in mind, we now consider the effect of narrowing or broadening a signal in the time domain by simple scaling of the time variable.

3. If  $c$  is a positive number, let  $f_c(x) = f(cx)$ . Show that

$$\hat{f}_c(\alpha) = \frac{1}{c} \hat{f}\left(\frac{\alpha}{c}\right).$$

Thus narrowing the signal function  $f$  in the time domain ( $c > 1$ ) broadens its transform in the frequency domain, and broadening the signal function  $f$  in the time domain ( $c < 1$ ) narrows its transform in the frequency domain.

To quantify the effect that we observe in Problem 3, we need to settle on a measure of the “width” of the graph of a function. The most commonly used measure is the root mean square width, which when applied to a signal  $f$  in the time and frequency domains yields a root mean square duration  $D(f)$  and a root mean square bandwidth  $B(f)$  given by

$$[D(f)]^2 = \frac{\int_{-\infty}^{\infty} x^2 [f(x)]^2 dx}{\int_{-\infty}^{\infty} [f(x)]^2 dx}$$

and

$$[B(f)]^2 = \frac{\int_{-\infty}^{\infty} \alpha^2 |\hat{f}(\alpha)|^2 d\alpha}{\int_{-\infty}^{\infty} |\hat{f}(\alpha)|^2 d\alpha}.$$

The bandwidth and duration are calculated relative to “centers” of  $\alpha = 0$  and  $x = 0$  because, by Problems 1 and 2, the graph of  $|\hat{f}(\alpha)|^2$  is symmetric around  $\alpha = 0$  in the frequency domain, and the signal can be shifted horizontally in the time domain without affecting the graph of  $|\hat{f}(\alpha)|^2$  in the frequency domain.

4. Show that for a family of functions  $f_c(x)$  defined in Problem 3,  $D(f_c) \cdot B(f_c)$  is independent of  $c$ .
5. Show that for the family of functions  $f_c(x) = e^{-c|x|}$ ,

$$D(f_c) \cdot B(f_c) = \frac{\sqrt{2}}{2}. \quad [\text{Hint: By Problem 4, you can just}$$

take  $f(x) = f_1(x)$ . The necessary Fourier integral can be gleaned quickly from Example 3 of Section 15.3. To evaluate the integrals in  $D(f)$  and  $B(f)$ , think about integration by parts and partial fractions, respectively.]

So the duration and the bandwidth of a signal are in a sense inversely proportional to each other under scaling of the time variable. What about the constant of proportionality? How small can  $D(f) \cdot B(f)$  be? Remarkably, there is a lower limit for this product.

6. Derive the **uncertainty inequality**: If

$$\int_{-\infty}^{\infty} [f(x)]^2 dx < \infty, \quad \int_{-\infty}^{\infty} |\hat{f}(\alpha)|^2 d\alpha < \infty,$$

and

$$\lim_{x \rightarrow \pm\infty} x[f(x)]^2 = 0,$$

then  $D(f) \cdot B(f) \geq \frac{1}{2}$ . Follow these steps.

(a) Establish Parseval's formula:

$$\int_{-\infty}^{\infty} [f(x)]^2 dx = \frac{1}{2\pi} \int_{-\infty}^{\infty} |\hat{f}(\alpha)|^2 d\alpha.$$

[Hint: Apply the convolution theorem given in Problem 20, Exercises 15.4 with  $g(x) = f(-x)$ .

Specifically, apply the formula for the inverse Fourier transform given in (6) of Section 15.4, show that  $\hat{g}(\alpha) = \hat{f}(\alpha)$ , and then let  $x = 0$ .]

(b) Establish the Schwartz inequality: For real-valued functions  $h_1$  and  $h_2$ ,

$$\left| \int_a^b h_1(s)h_2(s) ds \right|^2 \leq \left( \int_a^b [h_1(s)]^2 ds \right) \left( \int_a^b [h_2(s)]^2 ds \right)$$

with equality occurring only when  $h_2 = ch_1$ , where  $c$  is a constant [Hint: Write

$$\int_a^b [\lambda h_1(s) - h_2(s)]^2 ds$$

as a quadratic expression  $A\lambda^2 + B\lambda + C$  in the real variable  $\lambda$ . Note that the quadratic is nonnegative for all  $\lambda$  and consider the discriminant  $B^2 - 4AC$ .]

(c) Establish the uncertainty inequality. [Hint: First, apply the Schwartz inequality as follows:

$$\left| \int_{-\infty}^{\infty} xf(x)f'(x) dx \right|^2 \leq \left( \int_{-\infty}^{\infty} [xf(x)]^2 dx \right) \left( \int_{-\infty}^{\infty} [f'(x)]^2 dx \right).$$

Use integration by parts to show that

$$\int_{-\infty}^{\infty} xf(x)f'(x) dx = -\frac{1}{2} \int_{-\infty}^{\infty} [f(x)]^2 dx.$$

Rewrite the second integral appearing on the right-hand side of the inequality using the operational property (11) of Section 15.4 and Parseval's formula.]

7. (a) Show that if  $f$  gives the minimum possible value of  $D(f) \cdot B(f)$ , then

$$f'(x) = cx f(x)$$

where  $c$  is some constant. Solve this differential equation to show that  $f(x) = de^{cx^2/2}$  for  $c < 0$  and  $d$  is a constant. (Such a function is called a Gaussian function. Gaussian functions play an important role in probability theory.)

(b) Take the Fourier transform of both sides of the differential equation in part (a) to obtain a differential equation for  $\hat{f}(\alpha)$  and show that  $\hat{f}(\alpha) = \hat{f}(0)e^{i\alpha^2/(2c)}$ , where  $c$  is the same as in part (a). You will need the following fact:

$$\begin{aligned} \frac{d}{d\alpha} \hat{f}(\alpha) &= \frac{d}{d\alpha} \int_{-\infty}^{\infty} f(x)e^{i\alpha x} dx = \int_{-\infty}^{\infty} \frac{\partial}{\partial \alpha} f(x)e^{i\alpha x} dx \\ &= \int_{-\infty}^{\infty} ix f(x)e^{i\alpha x} dx = i\widehat{xf(x)}. \end{aligned}$$

(In Problem 35 in Exercises 9.11, we saw that  $\int_{-\infty}^{\infty} e^{-x^2} dx = \sqrt{\pi}$ . From this fact you can deduce that  $\hat{f}(0) = \sqrt{2\pi/|c|} \cdot d$ .)

So the minimum possible value of  $D(f) \cdot B(f)$  is attained for a Gaussian function, whose Fourier transform is another Gaussian function!

The word “uncertainty” is associated with the inequality presented in Problem 6 because, from a more abstract point of view, it is mathematically analogous to the famous Heisenberg uncertainty principle of quantum mechanics. (The interpretation of this principle of quantum mechanics is a subtle matter, but it is commonly understood as “the more accurately one determines the position of a particle, the less accurately one knows its momentum, and vice versa.”)